

# MCP6541/2/3/4

## **Push-Pull Output Sub-Microamp Comparators**

#### Features

- Low Quiescent Current: 600 nA/comparator (typ.)
- Rail-to-Rail Input: V<sub>SS</sub>-0.3V to V<sub>DD</sub>+0.3V
- CMOS/TTL Compatible Output
- Propagation Delay 4 µs (typ)
- Wide Supply Voltage Range: 1.6V to 5.5V
- Available in Single, Dual and Quad
- · Single available in SOT-23-5, SC-70-5 packages
- Chip Select (CS) with MCP6543
- · Low Switching Current
- Internal Hysteresis: 3.3 mV (typ)

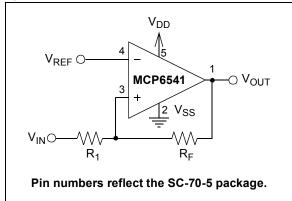
### **Typical Applications**

- Laptop Computers
- Mobile Phones
- · Metering Systems
- · Hand-held Electronics
- RC Timers
- · Alarm and Monitoring Circuits
- · Windowed Comparators
- Multi-vibrators

#### **Related Devices**

Open-Drain Output: MCP6546/7/8/9

#### **Typical Application Circuit**



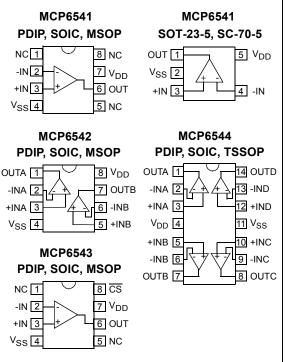
#### Description

The Microchip Technology, Inc. MCP6541/2/3/4 family of comparators is offered in single (MCP6541), single with chip select (MCP6543), dual (MCP6542) and quad (MCP6544) configurations. The outputs are push-pull (CMOS/TTL compatible) and are capable of driving heavy DC or capacitive loads.

These comparators are optimized for low power, single-supply operation with greater than rail-to-rail input operation. The push-pull output of the family supports rail-to-rail output swing and interfaces with TTL/CMOS logic. Input hysteresis eliminates output switching due to internal input noise voltage, reducing current draw. The output limits supply current surges and dynamic power consumption, while switching. This product family operates with a single-supply voltage as low as 1.6V and draws less than 1  $\mu$ A/comparator of quiescent current.

The related MCP6546/7/8/9 family of comparators from Microchip has an open-drain output. Used with a pullup resistor, these devices can be used as level-shifters for any desired voltage up to 10V and in wired-OR logic.

## Package Types



## 1.0 ELECTRICAL CHARACTERISTICS

#### 1.1 Absolute Maximum Ratings†

V <sub>DD</sub> - V <sub>SS</sub>	7.0V
All inputs and outputs $V_{SS}$	–0.3V to V <sub>DD</sub> +0.3V
Difference Input voltage	V <sub>DD</sub> - V <sub>SS</sub>
Output Short-Circuit Current	continuous
Current at Input Pins	±2 mA
Current at Output and Supply Pins	±30 mA
Storage temperature	65°C to +150°C
Junction temperature	+150°C
ESD protection on all pins (HBM); (MM).	4 kV; 400V

**†** Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

## **PIN FUNCTION TABLE**

NAME	FUNCTION
+IN/+INA/+INB/+INC/+IND	Non-inverting Inputs
-IN/-INA/-INB/-INC/-IND	Inverting Inputs
V <sub>DD</sub>	Positive Power Supply
V <sub>SS</sub>	Negative Power Supply
OUT/OUTA/OUTB/OUTC/OUTD	Outputs
CS	Chip Select
NC	Not Connected

## DC CHARACTERISTICS

**Electrical Specifications:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = 25°C,  $V_{IN+}$  =  $V_{DD}/2$ ,  $V_{IN-}$  =  $V_{SS}$ , and  $R_L$  = 100 k $\Omega$  to  $V_{DD}/2$ . Refer to Figure 1-3.

Parameters	Sym	Min	Тур	Мах	Units	Conditions
Power Supply						
Supply Voltage	V <sub>DD</sub>	1.6	—	5.5	V	
Quiescent Current per comparator	Ι <sub>Q</sub>	0.3	0.6	1.0	μA	I <sub>OUT</sub> = 0
Input						
Input Voltage Range	V <sub>CMR</sub>	V <sub>SS</sub> -0.3	—	V <sub>DD</sub> +0.3	V	
Common Mode Rejection Ratio	CMRR	55	70	—	dB	$V_{DD}$ = 5V, $V_{CM}$ = -0.3V to 5.3V
Common Mode Rejection Ratio	CMRR	50	65	—	dB	$V_{DD}$ = 5V, $V_{CM}$ = 2.5V to 5.3V
Common Mode Rejection Ratio	CMRR	55	70	—	dB	$V_{DD}$ = 5V, $V_{CM}$ = -0.3V to 2.5V
Power Supply Rejection Ratio	PSRR	63	80	—	dB	$V_{CM} = V_{SS}$
Input Offset Voltage	V <sub>OS</sub>	-7.0	±1.5	+7.0	mV	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Drift with Temperature	$\Delta V_{OS} / \Delta T$	—	±3	—	µV/°C	$T_A = -40^{\circ}C$ to +85°C, $V_{CM} = V_{SS}$
Input Hysteresis Voltage	V <sub>HYST</sub>	1.5	3.3	6.5	mV	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Drift with Temperature	$\Delta V_{HYST} / \Delta T$	—	10	—	µV/°C	$T_A = -40^{\circ}C$ to +25°C, $V_{CM} = V_{SS}$
Drift with Temperature	$\Delta V_{HYST} / \Delta T$	—	5	—	µV/°C	$T_A$ = +25°C to +85°C, $V_{CM}$ = $V_{SS}$
Input Bias Current	I <sub>B</sub>	—	1	—	pА	$V_{CM} = V_{SS}$
Over Temperature	I <sub>B</sub>	—	—	100	pА	$T_A = -40^{\circ}C$ to +85°C, $V_{CM} = V_{SS}$ (Note 3)
Input Offset Current	I <sub>OS</sub>	—	±1	—	pА	$V_{CM} = V_{SS}$
Common Mode Input Impedance	Z <sub>CM</sub>	—	10 <sup>13</sup>   4	—	Ω  pF	
Differential Input Impedance	Z <sub>DIFF</sub>	—	10 <sup>13</sup>   2	—	Ω  pF	
Push Pull Output						
High-Level Output Voltage	V <sub>OH</sub>	V <sub>DD</sub> -0.2	—	—	V	I <sub>OUT</sub> = -2 mA, V <sub>DD</sub> = 5V
Low-Level Output Voltage	V <sub>OL</sub>	—	—	V <sub>SS</sub> +0.2	V	I <sub>OUT</sub> = 2 mA, V <sub>DD</sub> = 5V
Short Circuit Current	I <sub>SC</sub>	_	±50	—	mA	(Note 2)

**Note** 1: The input offset voltage is the center (average) of the input-referred trip points. The input hysteresis is the difference between the input-referred trip points.

2: Limit the output current to Absolute Maximum Rating of 30 mA.

3: Input bias current over temperature is not tested for SC-70-5 package.

## **AC CHARACTERISTICS**

**Electrical Specifications:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = 25°C,  $V_{IN+}$  =  $V_{DD}/2$ , Step = 200 mV, Overdrive = 100 mV, and  $C_L$  = 36 pF. Refer to Figure 1-2 and Figure 1-3.

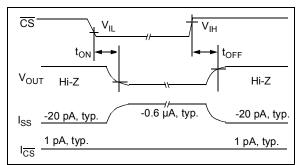
Parameters	Sym	Min	Тур	Мах	Units	Conditions
Rise Time	t <sub>R</sub>	_	0.85	_	μs	
Fall Time	t <sub>F</sub>	—	0.85	—	μs	
Propagation Delay (High-to-Low)	t <sub>PHL</sub>	_	4	8	μs	
Propagation Delay (Low-to-High)	t <sub>PLH</sub>	—	4	8	μs	
Propagation Delay Skew	t <sub>PDS</sub>	—	±0.2	—	μs	(Note 1)
Maximum Toggle Frequency	f <sub>MAX</sub>	_	160	_	kHz	V <sub>DD</sub> = 1.6V
	f <sub>MAX</sub>	—	120	-	kHz	V <sub>DD</sub> = 5.5V
Input Noise Voltage	E <sub>N</sub>	_	200	_	μV <sub>P-P</sub>	10 Hz to 100 kHz

**Note** 1: Propagation Delay Skew is defined as:  $t_{PDS} = t_{PLH} - t_{PHL}$ .

## SPECIFICATIONS FOR MCP6543 CHIP SELECT

**Electrical Specifications:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = 25°C,  $V_{IN+}$  =  $V_{DD}/2$ ,  $V_{IN-}$  =  $V_{SS}$ , and  $C_L$  = 36 pF. Refer to Figure 1-3.

Parameters	Sym	Min	Тур	Max	Units	Conditions
CS Low Specifications						
CS Logic Threshold, Low	V <sub>IL</sub>	$V_{SS}$	—	$0.2V_{\text{DD}}$	V	
CS Input Current, Low	I <sub>CSL</sub>	—	5.0	—	pА	$\overline{\text{CS}} = \text{V}_{\text{SS}}$
CS High Specifications						
CS Logic Threshold, High	VIH	0.8V <sub>DD</sub>	—	$V_{DD}$	V	
CS Input Current, High	I <sub>CSH</sub>	—	1	—	pА	$\overline{CS} = V_{DD}$
CS Input High, V <sub>DD</sub> Current	I <sub>DD</sub>	—	18	—	pА	$\overline{\text{CS}} = \text{V}_{\text{DD}}$
CS Input High, GND Current	I <sub>SS</sub>	—	-20	—	pА	$\overline{\text{CS}} = V_{\text{DD}}$
Comparator Output Leakage	I <sub>O(LEAK)</sub>	—	1	—	pА	V <sub>OUT</sub> = V <sub>DD</sub>
CS Dynamic Specifications						
CS Low to Comparator Output Low Turn-on Time	t <sub>ON</sub>	—	2	50	ms	$\overline{\text{CS}} = 0.2 \text{V}_{\text{DD}}$ to $\text{V}_{\text{OUT}} = \text{V}_{\text{DD}}/2$ , $\text{V}_{\text{IN-}} = \text{V}_{\text{DD}}$
CS High to Comparator Output High Z Turn-off Time	t <sub>OFF</sub>	—	10	_	μs	$\overline{CS} = 0.8V_{DD}$ to $V_{OUT} = V_{DD}/2$ , $V_{IN-} = V_{DD}$
CS Hysteresis	V <sub>CS_HYST</sub>	_	0.6	—	V	V <sub>DD</sub> = 5V



**FIGURE 1-1:** Timing Diagram for the  $\overline{CS}$  Pin on the MCP6543.

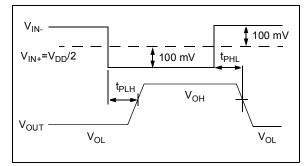
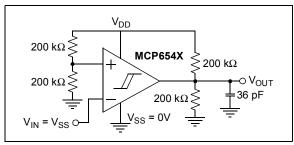


FIGURE 1-2: Propagation Delay Timing Diagram.

## **TEMPERATURE SPECIFICATIONS**

Electrical Specifications: Unless otherwise indicated, $V_{DD}$ = +1.6V to +5.5V and $V_{SS}$ = GND.									
Parameters	Sym	Min	Тур	Мах	Units	Conditions			
Temperature Ranges									
Specified Temperature Range	T <sub>A</sub>	-40	_	+85	°C				
Operating Temperature Range	T <sub>A</sub>	-40	_	+125	°C	Note			
Storage Temperature Range	T <sub>A</sub>	-65	_	+150	°C				
Thermal Package Resistances	i								
Thermal Resistance, 5L-SC-70	$\theta_{JA}$	_	331	_	°C/W				
Thermal Resistance, 5L-SOT-23	$\theta_{JA}$	—	256	_	°C/W				
Thermal Resistance, 8L-PDIP	$\theta_{JA}$	_	85	_	°C/W				
Thermal Resistance, 8L-SOIC	$\theta_{JA}$	_	163	_	°C/W				
Thermal Resistance, 8L-MSOP	$\theta_{JA}$	_	206	—	°C/W				
Thermal Resistance, 14L-PDIP	$\theta_{JA}$	—	70	—	°C/W				
Thermal Resistance, 14L-SOIC	$\theta_{JA}$	_	120	—	°C/W				
Thermal Resistance, 14L-TSSOP	$\theta_{JA}$	—	100	—	°C/W				

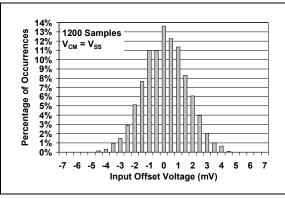
**Note:** The MCP6541/2/3/4 operates over this extended temperature range, but with reduced performance. In any case, the Junction Temperature  $(T_J)$  must not exceed the Absolute Maximum specification of +150°C.



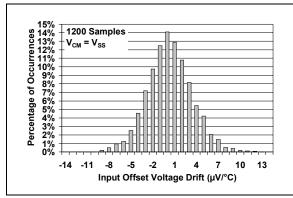
**FIGURE 1-3:** DC Test Circuit for the Push-Pull Output Comparators.

## 2.0 TYPICAL PERFORMANCE CURVES

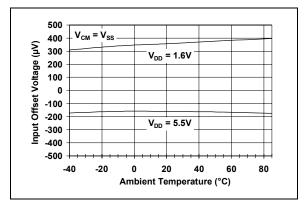
**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.



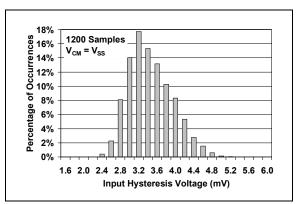
**FIGURE 2-1:** Histogram of Input Offset Voltage with  $V_{CM} = V_{SS}$ .



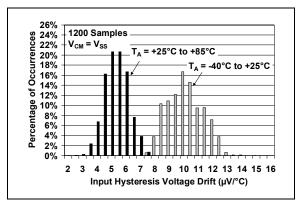
**FIGURE 2-2:** Histogram of Input Offset Voltage Drift with  $V_{CM} = V_{SS}$ .



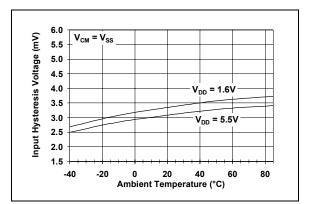
**FIGURE 2-3:** Input Offset Voltage vs. Ambient Temperature vs. Power Supply Voltage with  $V_{CM} = V_{SS}$ .



**FIGURE 2-4:** Histogram of Input Hysteresis Voltage with  $V_{CM} = V_{SS}$ .

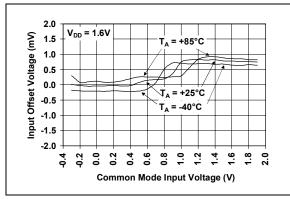


**FIGURE 2-5:** Histogram of Input Hysteresis Voltage Drift with Temperature =  $-40^{\circ}$ C to 25°C and 25°C to 85°C,  $V_{CM} = V_{SS}$ .

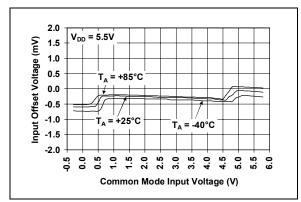


**FIGURE 2-6:** Input Hysteresis Voltage vs. Temperature vs. Power Supply Voltage with  $V_{CM} = V_{SS}$ .

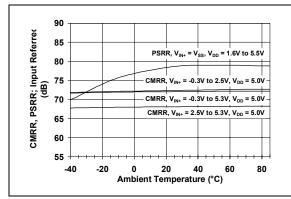
## MCP6541/2/3/4



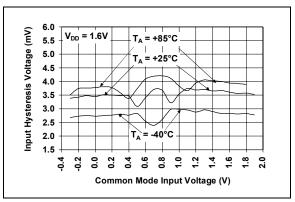
**FIGURE 2-7:** Input Offset Voltage vs. Common Mode Input Voltage vs. Temperature with  $V_{DD} = 1.6V$ .



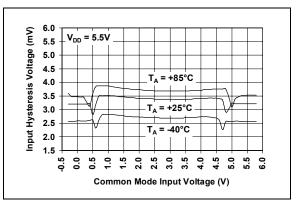
**FIGURE 2-8:** Input Offset Voltage vs. Common Mode Input Voltage vs. Temperature with  $V_{DD} = 5.5V$ .



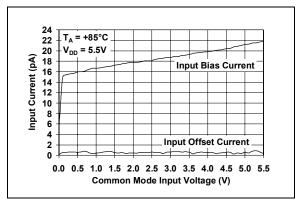
**FIGURE 2-9:** Common Mode Rejection Ratio, Power Supply Rejection Ratio vs. Ambient Temperature with  $V_{CM} = V_{SS}$ .



**FIGURE 2-10:** Input Hysteresis Voltage vs. Common Mode Input Voltage vs. Temperature with  $V_{DD} = 1.6V$ .



**FIGURE 2-11:** Input Hysteresis Voltage vs. Common Mode Input Voltage vs. Temperature with  $V_{DD} = 5.5V$ .



**FIGURE 2-12:** Input Bias Current, Input Offset Current vs. Common Mode Voltage with Temperature = 85°C.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +5.0V,  $V_{SS}$  = GND,  $T_A$  = 25°C,  $V_{IN+} = V_{DD}/2$ ,  $V_{IN-}$  = GND,  $R_L$  = 100 k $\Omega$  to  $V_{DD}/2$ , and  $C_L$  = 36 pF.

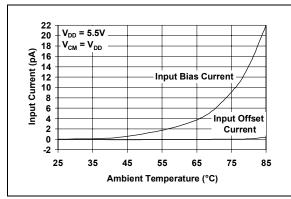


FIGURE 2-13: Input Bias Current, Input Offset Current vs. Ambient Temperature.

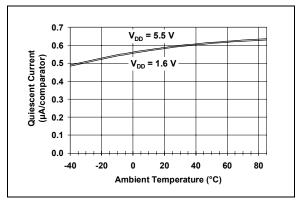


FIGURE 2-14: Quiescent Current vs. Ambient Temperature vs. Power Supply Voltage.

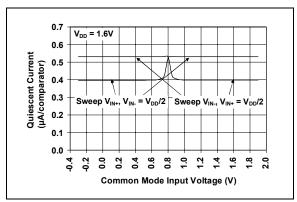


FIGURE 2-15:Quiescent Current vs.Common Mode Input Voltage with  $V_{DD} = 1.6V$ .

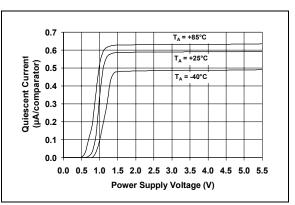
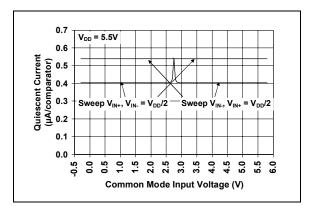
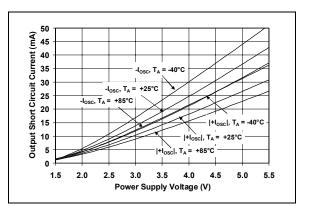


FIGURE 2-16: Quiescent Current vs. Power Supply Voltage vs. Temperature.

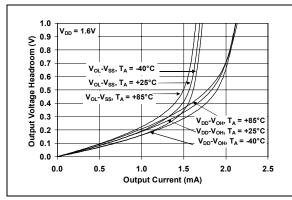


**FIGURE 2-17:** Quiescent Current vs. Common Mode Input Voltage with  $V_{DD} = 5V$ .

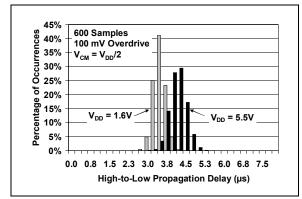


**FIGURE 2-18:** Output Short-Circuit Current vs. Power Supply Voltage vs. Temperature.

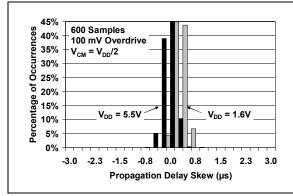
## MCP6541/2/3/4



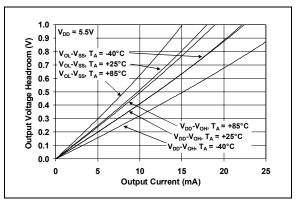
**FIGURE 2-19:** Output Voltage Headroom vs. Output Current vs. Temperature with  $V_{DD} = 1.6V$ .



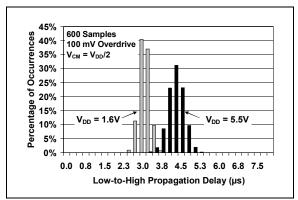
**FIGURE 2-20:** Histogram of High-to-Low Propagation Delay with  $V_{DD}$  = 1.6V and 5.5V.



**FIGURE 2-21:** Histogram of Propagation Delay Skew with  $V_{DD} = 1.6V$  and 5.5V.



**FIGURE 2-22:** Output Voltage Headroom vs. Output Current vs. Temperature with  $V_{DD} = 5.5V$ .



**FIGURE 2-23:** Histogram of Low-to-High Propagation Delay with  $V_{DD}$  = 1.6V and 5.5V.

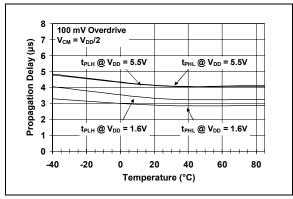


FIGURE 2-24:Propagation Delay vs.Temperature vs. Power Supply Voltage.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +5.0V,  $V_{SS}$  = GND,  $T_A$  = 25°C,  $V_{IN+} = V_{DD}/2$ ,  $V_{IN-}$  = GND,  $R_L$  = 100 k $\Omega$  to  $V_{DD}/2$ , and  $C_L$  = 36 pF.

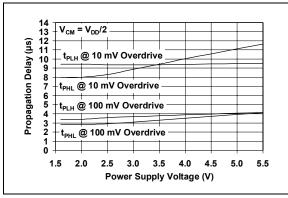


FIGURE 2-25:Propagation Delay vs.Power Supply Voltage vs. Input Overdrive.

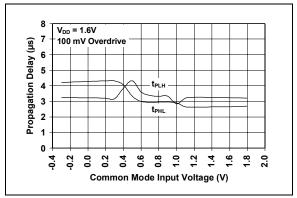


FIGURE 2-26:Propagation Delay vs.Common Mode Input Voltage with  $V_{DD} = 1.6V.$ 

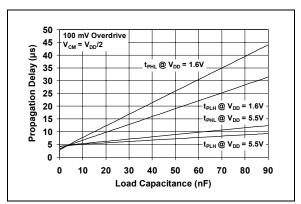


FIGURE 2-27: Propagation Delay vs. Load Capacitance vs. Power Supply Voltage.

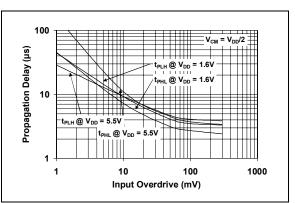
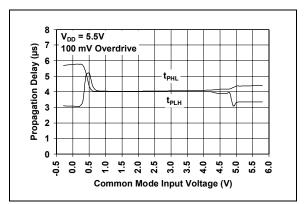


FIGURE 2-28: Propagation Delay vs. Input Overdrive vs. Power Supply Voltage.



**FIGURE 2-29:** Propagation Delay vs. Common Mode Input Voltage with  $V_{DD}$  = 5.5V.

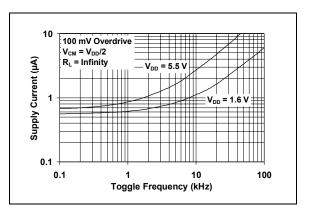


FIGURE 2-30: Supply Current vs. Toggle Frequency vs. Power Supply Voltage.

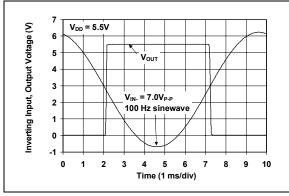
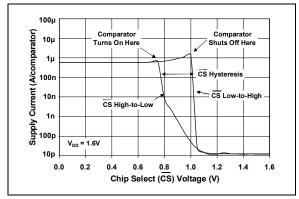
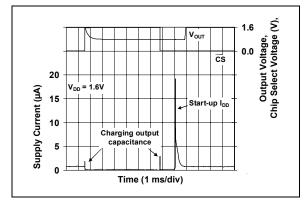


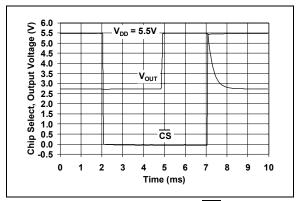
FIGURE 2-31:The MCP6541/2/3/4comparators show no phase reversal.



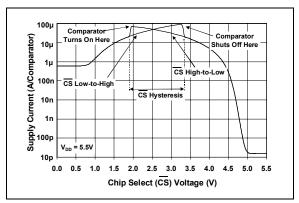
**FIGURE 2-32:** Supply Current (shoot through current) vs. Chip Select ( $\overline{CS}$ ) Voltage with  $V_{DD}$  = 1.6V (MCP6543 only).



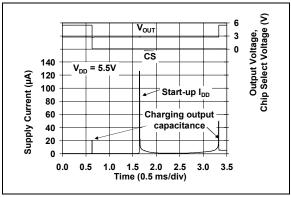
**FIGURE 2-33:** Supply Current (charging current) vs. Chip Select (CS) pulse with  $V_{DD} = 1.6V$  (MCP6543 only).



**FIGURE 2-34:** Chip Select (CS) Step Response (MCP6543 only).



**FIGURE 2-35:** Supply Current (shoot through current) vs. Chip Select (CS) Voltage with  $V_{DD} = 5.5V$  (MCP6543 only).



**FIGURE 2-36:** Supply Current (charging current) vs. Chip Select ( $\overline{CS}$ ) pulse with  $V_{DD}$  = 5.5V (MCP6543 only).

## 3.0 APPLICATIONS INFORMATION

The MCP6541/2/3/4 family of push-pull, output comparators are fabricated on Microchip's state-of-the-art CMOS process. They are suitable for a wide range of applications requiring very low power consumption. A 0.1  $\mu$ F bypass capacitor from the power supply pin to ground is recommended.

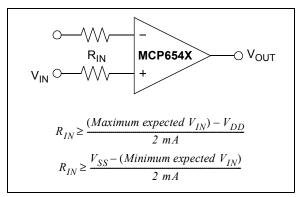
#### 3.1 Rail-to-Rail Input

The input stage of this family of devices uses two differential input stages in parallel: one operates at low input voltages and the other at high input voltages. With this topology, the input voltage is +0.3V above  $V_{DD}$  and -0.3V below  $V_{SS}$ . The input offset voltage is measured at both  $V_{SS}$  - 0.3V and  $V_{DD}$  + 0.3V to ensure proper operation.

#### 3.2 Input Voltage and Phase Reversal

The comparator family uses CMOS transistors at the input. They are designed to not exhibit phase inversion when the input pins exceed the supply voltages. Figure 2-31 shows an input voltage exceeding both supplies with no resulting phase inversion.

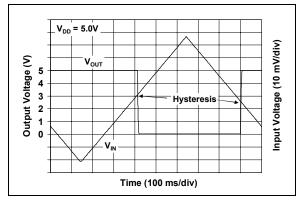
The maximum operating input voltages that can be applied are  $V_{SS}$ -0.3V and  $V_{DD}$ +0.3V. Voltages on the inputs that exceed this absolute maximum rating can cause excessive current to flow and permanently damage the device. In applications where the input pin exceeds the specified range, external resistors can be used to limit the current below ±2 mA, as shown in Figure 3-1.



**FIGURE 3-1:** An input resistor (*R<sub>IN</sub>*) should be used to limit excessive input current if either of the inputs exceeds the Absolute Maximum specification.

#### 3.3 Hysteresis

Input offset voltage ( $V_{OS}$ ) is the center (average) of the (input-referred) low-high and high-low trip points. Input hysteresis voltage ( $V_{HYST}$ ) is the difference between the same trip points. Hysteresis reduces output chattering when one input is slowly moving past the other and thus reduces dynamic supply current. It also helps in systems where it is best not to cycle between states too frequently (e.g., air conditioner thermostatic control). The family has internally set hysteresis which is small enough to maintain input offset accuracy (<7 mV) and large enough to eliminate output chattering caused by the comparator's own input noise voltage (200  $\mu$ Vp-p).



**FIGURE 3-2:** The MCP6541/2/3/4 comparators' internal hysteresis eliminates output chatter caused by input noise voltage.

#### 3.3.1 NON-INVERTING CIRCUIT

Greater flexibility in selecting hysteresis, or input trip points, is achieved by using external resistors. Figure 3-3 shows a non-inverting circuit for single-supply applications using just two resistors. The resulting hysteresis diagram is shown in Figure 3-4.

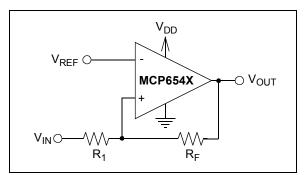
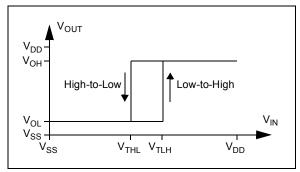


FIGURE 3-3: Non-inverting circuit with hysteresis for single-supply.



**FIGURE 3-4:** Hysteresis Diagram for the Non-Inverting Circuit.

The trip points for Figures 3-3 and 3-4 are:

#### EQUATION

$$V_{TLH} = V_{REF} \left( 1 + \frac{R_I}{R_F} \right) - V_{OL} \left( \frac{R_I}{R_F} \right)$$
$$V_{THL} = V_{REF} \left( 1 + \frac{R_I}{R_F} \right) - V_{OH} \left( \frac{R_I}{R_F} \right)$$
$$V_{TLH} = trip \ voltage \ from \ low \ to \ high$$
$$V_{THL} = trip \ voltage \ from \ high \ to \ low$$

#### 3.3.2 INVERTING CIRCUIT

Figure 3-5 shows an inverting circuit for single-supply using three resistors. The resulting hysteresis diagram is shown in Figure 3-6.

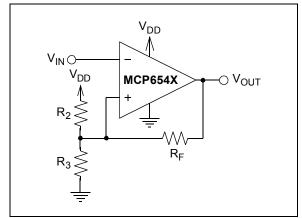
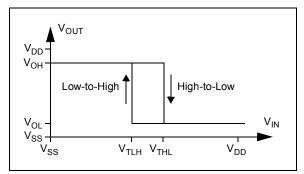


FIGURE 3-5: Inverting Circuit With Hysteresis.



**FIGURE 3-6:** Hysteresis Diagram for the Inverting Circuit.

The trip points for Figures 3-5 and 3-6 are given by:

#### EQUATION

$$V_{REF} = V_{DD} \times \frac{R_3}{R_2 + R_3}$$

$$R_{23} = \frac{R_2 R_3}{R_2 + R_3}$$

$$V_{THL} = V_{OH} \left(\frac{R_{23}}{R_{23} + R_F}\right) + V_{REF} \left(\frac{R_F}{R_{23} + R_F}\right)$$

$$V_{TLH} = V_{OL} \left(\frac{R_{23}}{R_{23} + R_F}\right) + V_{REF} \left(\frac{R_F}{R_{23} + R_F}\right)$$

### 3.4 The MCP6543 Chip Select (CS) Option

The MCP6543 is a single comparator with a chip select (CS) option. When CS is pulled high, the total current consumption drops to 20 pA (typ); 1 pA (typ) flows through the CS pin, 1 pA (typ) flows through the output pin and 18 pA (typ) flows through the V<sub>DD</sub> pin as shown in Figure 1-1. When this happens, the comparator <u>output</u> is put into a high-impedance state. By pulling CS low, the comparator will not operate properly. Figure 1-1 shows the <u>output</u> voltage and supply current response to a CS pulse.

The internal  $\overline{CS}$  circuitry is designed to minimize glitches when cycling the  $\overline{CS}$  pin. This helps conserve power, which is especially important in battery-powered applications.

#### 3.5 Push-Pull Output

The push-pull output is designed to be compatible with CMOS and TTL logic. The output transistors are configured to give rail-to-rail output performance. They are driven with circuitry that minimizes any switching current (shoot-through current from supply-to-supply) when the output is transitioned from high-to-low, or from low-to-high. See Figures 2-15, 2-17, 2-32 through 2-36 for more information.

#### 3.6 Capacitive Loads

Reasonable capacitive loads (e.g., logic gates) have little impact on propagation delay (see Figure 2-27). The supply current increases with increasing toggle frequency (Figure 2-30), especially with higher capacitive loads.

#### 3.7 Battery Life

In order to maximize battery life in portable applications, use large resistors and small capacitive loads. Also, avoid toggling the out<u>put</u> more than necessary, and do not use chip select  $(\overline{CS})$  to conserve power for short periods of time. Capacitive loads will draw additional power at start-up.

#### 3.8 Layout Considerations

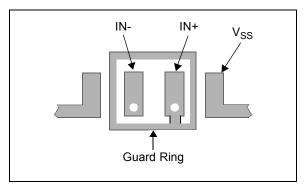
Good PC board layout techniques will help you achieve the performance shown in the specs and Typical Performance Curves. It will also help you minimize EMC (Electro-Magnetic Compatibility) issues.

#### 3.8.1 SURFACE LEAKAGE

In applications where low input bias current is critical, PC board surface leakage effects and signal coupling from trace-to-trace need to be considered.

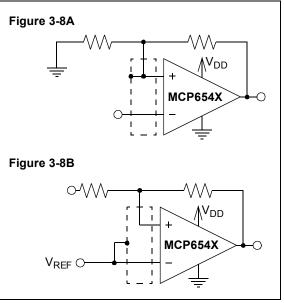
Surface leakage is caused by a difference in voltage between traces, combined with high humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is  $10^{12}\Omega$ . A 5V difference would cause 5 pA of current to flow, which is greater than the input current of the family at 25°C (1 pA, typ).

The simplest technique to reduce surface leakage is using a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin or trace. Figure 3-7 shows an example of a typical layout.



## FIGURE 3-7: Example of Guard Ring Layout.

Circuit schematics for different guard ring implementations are shown in Figure 3-8. Figure 3-8A biases the guard ring to the input common mode voltage. Figure 3-8B biases the guard ring to a reference voltage ( $V_{REF}$ , which can be ground). Place the guard ring on the node that is the most constant.



**FIGURE 3-8:** Two Possible Guard Ring Connection Strategies To Reduce Surface Leakage Effects.

#### 3.8.2 COMPONENT PLACEMENT

Separate digital from analog and low-speed from highspeed. This helps prevent crosstalk.

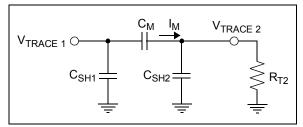
Keep sensitive traces short and straight. Separate them from interfering components and traces. This is especially important for high frequency (low rise time) signals.

Use a 0.1  $\mu F$  supply bypass capacitor within 0.1" (2.5 mm) of the  $V_{DD}$  pin. It must connect directly to the ground plane.

#### 3.8.3 SIGNAL COUPLING

The input pins of the MCP6541/2/3/4 family of comparators are high-impedance, which allows noise injection. This noise can be capacitively or magnetically coupled. In either case, using a ground plane helps reduce noise injection.

When noise is coupled capacitively, ground plane reduces the coupling capacitance and provides shunt capacitance to ground for high-frequency signals; Figure 3-8 shows the equivalent circuit. The coupled current ( $I_M$ ) produces a lower voltage ( $V_{TRACE 2}$ ) on the victim trace when the trace-to-ground plane capacitance ( $C_{SH2}$ ) is large, and the terminating resistor ( $R_{T2}$ ) is small. Increasing the distance between traces, and using wider traces, also helps.



**FIGURE 3-9:** Equivalent Circuit for Capacitive Coupling Between Traces on a PC Board (with Ground Plane).

When noise is coupled magnetically, a ground plane reduces the mutual inductance between traces. This occurs because the ground return current at high frequencies will follow a path directly beneath the signal trace. Increasing the separation between traces makes a significant difference. Changing the direction of one of the traces can also reduce magnetic coupling.

If these techniques are not enough, it may help to place guard traces next to the victim trace. They should be on both sides of the victim trace, and as close as possible. Connect the guard traces to ground plane at both ends, and in the middle for long traces.

#### 3.9 Typical Applications

#### 3.9.1 PRECISE COMPARATOR

Some applications require higher DC precision. An easy way to solve this problem is to use an amplifier to gain up the input signal before it reaches the comparator. Figure 3-10 shows an example of this approach.

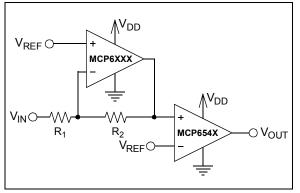


FIGURE 3-10: Precise Inverting Comparator.

#### 3.9.2 WINDOWED COMPARATOR

Figure 3-11 shows one approach to designing a windowed comparator. The AND gate produces a logic 1 when the input voltage is between  $V_{RB}$  and  $V_{RT}$  (where  $V_{RT} > V_{RB}$ ).

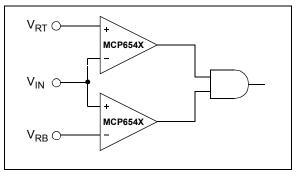


FIGURE 3-11: Windowed Comparator.

#### 3.9.3 BISTABLE MULTI-VIBRATOR

A simple bistable multi-vibrator design is shown in Figure 3-12.  $V_{REF}$  needs to be between the power supplies ( $V_{SS}$  = GND and  $V_{DD}$ ) to achieve oscillation. The output duty cycle changes with  $V_{REF}$ .

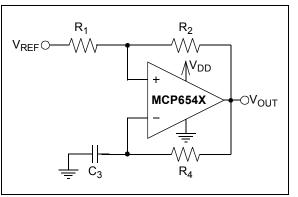
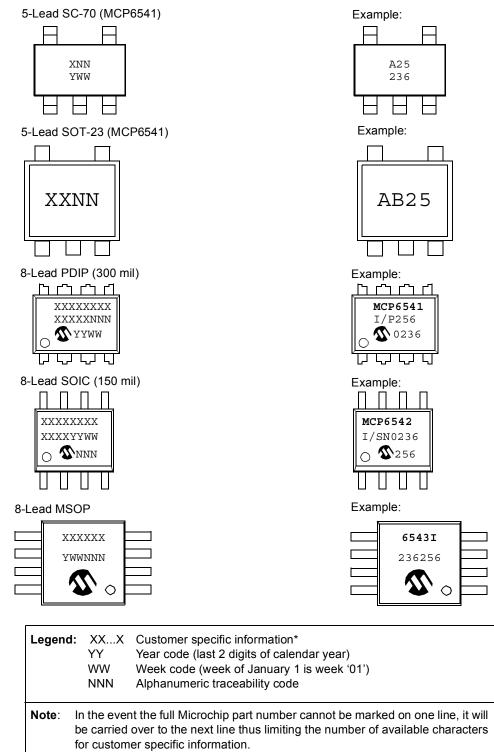


FIGURE 3-12: Bistable Multi-vibrator.

### 4.0 PACKAGING INFORMATION

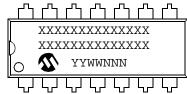
#### 4.1 Package Marking Information



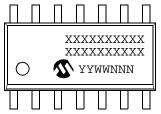
\* Standard marking consists of Microchip part number, year code, week code, traceability code (facility code, mask rev#, and assembly code). For marking beyond this, certain price adders apply. Please check with your Microchip Sales Office.

### Package Marking Information (Continued)

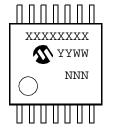
14-Lead PDIP (300 mil) (MCP6544)



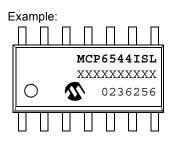
14-Lead SOIC (150 mil) (MCP6544)



14-Lead TSSOP (MCP6544)



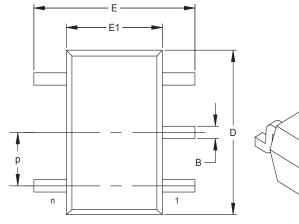
Example: MCP6544-I/P XXXXXXXXXXXX 0236256 UUUUUUUUUU

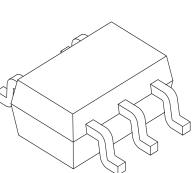


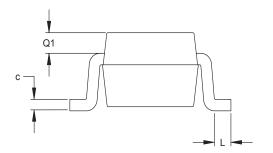
Example:

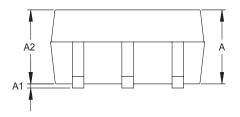


5-Lead Plastic Package (LT) (SC-70)









	Units	INCHES			М	ILLIMETERS	*
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5			5	
Pitch	р		.026 (BSC)			0.65 (BSC)	
Overall Height	А	.031		.043	0.80		1.10
Molded Package Thickness	A2	.031		.039	0.80		1.00
Standoff	A1	.000		.004	0.00		0.10
Overall Width	E	.071		.094	1.80		2.40
Molded Package Width	E1	.045		.053	1.15		1.35
Overall Length	D	.071		.087	1.80		2.20
Foot Length	L	.004		.012	0.10		0.30
Top of Molded Pkg to Lead Shoulder	Q1	.004		.016	0.10		0.40
Lead Thickness	С	.004		.007	0.10		0.18
Lead Width	В	.006		.012	0.15		0.30

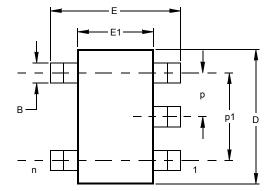
\*Controlling Parameter

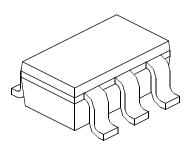
Notes:

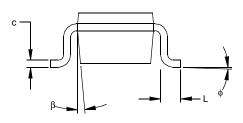
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. JEITA (EIAJ) Standard: SC-70

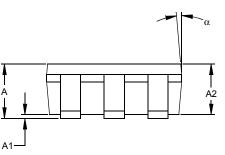
Drawing No. C04-061

## 5-Lead Plastic Small Outline Transistor (OT) (SOT23)









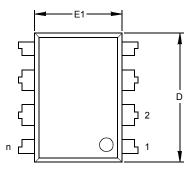
	Units		INCHES*			<b>IILLIMETERS</b>	5
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5			5	
Pitch	р		.038			0.95	
Outside lead pitch (basic)	p1		.075			1.90	
Overall Height	Α	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff §	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	φ	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

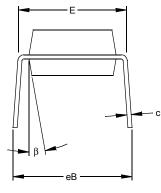
\* Controlling Parameter § Significant Characteristic

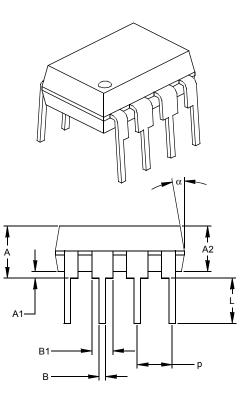
Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MO-178 Drawing No. C04-091

## 8-Lead Plastic Dual In-line (P) – 300 mil (PDIP)







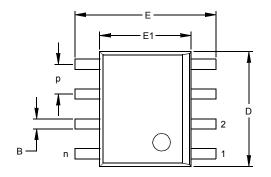
	Units		INCHES*		MILLIMETERS		
Dimensio	n Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р		.100			2.54	
Top to Seating Plane	А	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.360	.373	.385	9.14	9.46	9.78
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

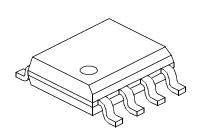
\* Controlling Parameter § Significant Characteristic

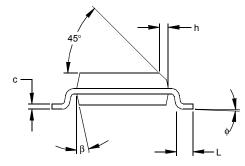
Notes: Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-001

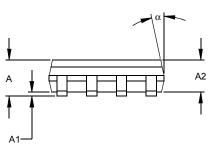
Drawing No. C04-018

### 8-Lead Plastic Small Outline (SN) – Narrow, 150 mil (SOIC)









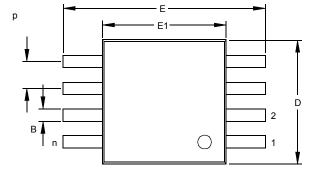
	Units		INCHES*		N	MILLIMETERS		
Dimensio	n Limits	MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n		8			8		
Pitch	р		.050			1.27		
Overall Height	Α	.053	.061	.069	1.35	1.55	1.75	
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55	
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25	
Overall Width	Е	.228	.237	.244	5.79	6.02	6.20	
Molded Package Width	E1	.146	.154	.157	3.71	3.91	3.99	
Overall Length	D	.189	.193	.197	4.80	4.90	5.00	
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51	
Foot Length	L	.019	.025	.030	0.48	0.62	0.76	
Foot Angle	ф	0	4	8	0	4	8	
Lead Thickness	С	.008	.009	.010	0.20	0.23	0.25	
Lead Width	В	.013	.017	.020	0.33	0.42	0.51	
Mold Draft Angle Top	α	0	12	15	0	12	15	
Mold Draft Angle Bottom	β	0	12	15	0	12	15	
• ·		-		-	-			

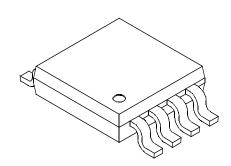
\* Controlling Parameter § Significant Characteristic

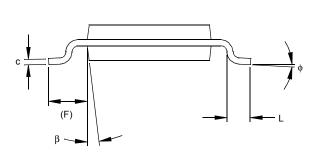
Notes:

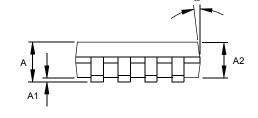
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-012 Drawing No. C04-057

## 8-Lead Plastic Micro Small Outline Package (MS) (MSOP)









	Units		INCHES		M		
Dimens	sion Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8				8
Pitch	р		.026			0.65	
Overall Height	А			.044			1.18
Molded Package Thickness	A2	.030	.034	.038	0.76	0.86	0.97
Standoff §	A1	.002		.006	0.05		0.15
Overall Width	E	.184	.193	.200	4.67	4.90	.5.08
Molded Package Width	E1	.114	.118	.122	2.90	3.00	3.10
Overall Length	D	.114	.118	.122	2.90	3.00	3.10
Foot Length	L	.016	.022	.028	0.40	0.55	0.70
Footprint (Reference)	F	.035	.037	.039	0.90	0.95	1.00
Foot Angle	φ	0		6	0		6
Lead Thickness	С	.004	.006	.008	0.10	0.15	0.20
Lead Width	В	.010	.012	.016	0.25	0.30	0.40
Mold Draft Angle Top	α		7			7	
Mold Draft Angle Bottom	β		7			7	

\*Controlling Parameter § Significant Characteristic

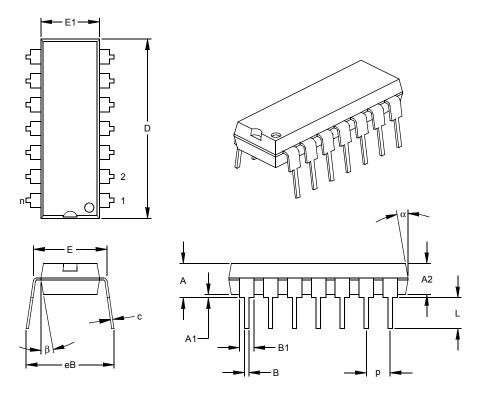
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Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.

Drawing No. C04-111

14-Lead Plastic Dual In-line (P) – 300 mil (PDIP)



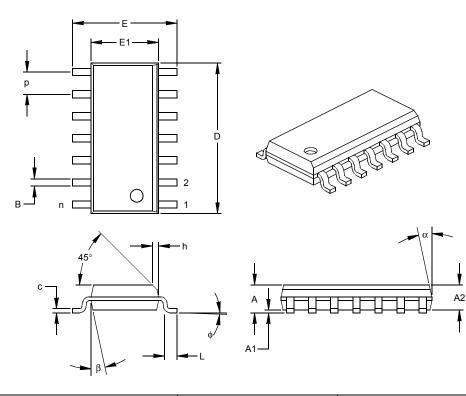
	Units		INCHES*		MILLIMETERS			
Dimensi	Dimension Limits			MAX	MIN	NOM	MAX	
Number of Pins	n		14			14		
Pitch	р		.100			2.54		
Top to Seating Plane	А	.140	.155	.170	3.56	3.94	4.32	
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68	
Base to Seating Plane	A1	.015			0.38			
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26	
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60	
Overall Length	D	.740	.750	.760	18.80	19.05	19.30	
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43	
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38	
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78	
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56	
Overall Row Spacing	§ eB	.310	.370	.430	7.87	9.40	10.92	
Mold Draft Angle Top	α	5	10	15	5	10	15	
Mold Draft Angle Bottom	β	5	10	15	5	10	15	

\* Controlling Parameter § Significant Characteristic

Notes: Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

.010" (0.254mm) per side. JEDEC Equivalent: MS-001 Drawing No. C04-005

## 14-Lead Plastic Small Outline (SL) - Narrow, 150 mil (SOIC)



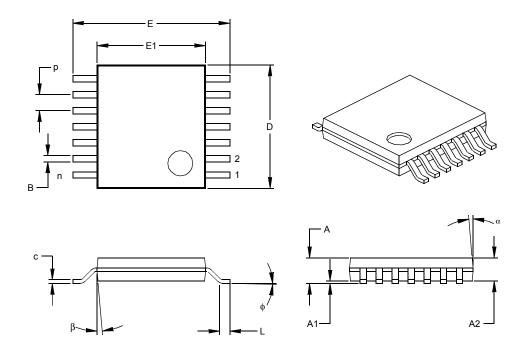
	Units		INCHES*		MILL		
Dimens	sion Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		14			14	
Pitch	р		.050			1.27	
Overall Height	Α	.053	.061	.069	1.35	1.55	1.75
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25
Overall Width	E	.228	.236	.244	5.79	5.99	6.20
Molded Package Width	E1	.150	.154	.157	3.81	3.90	3.99
Overall Length	D	.337	.342	.347	8.56	8.69	8.81
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51
Foot Length	L	.016	.033	.050	0.41	0.84	1.27
Foot Angle	¢	0	4	8	0	4	8
Lead Thickness	С	.008	.009	.010	0.20	0.23	0.25
Lead Width	В	.014	.017	.020	0.36	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15

\* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-012 Drawing No. C04-065

14-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm (TSSOP)



	Units	INCHES		MILLIMETERS*			
Dimensio	on Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		14			14	
Pitch	р		.026			0.65	
Overall Height	А			.043			1.10
Molded Package Thickness	A2	.033	.035	.037	0.85	0.90	0.95
Standoff §	A1	.002	.004	.006	0.05	0.10	0.15
Overall Width	Е	.246	.251	.256	6.25	6.38	6.50
Molded Package Width	E1	.169	.173	.177	4.30	4.40	4.50
Molded Package Length	D	.193	.197	.201	4.90	5.00	5.10
Foot Length	L	.020	.024	.028	0.50	0.60	0.70
Foot Angle	φ	0	4	8	0	4	8
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	B1	.007	.010	.012	0.19	0.25	0.30
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

\* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. JEDEC Equivalent: MO-153 Drawing No. C04-087

## **PRODUCT IDENTIFICATION SYSTEM**

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	- <u>x</u> /xx		Ex	amples:	
Device	Temperature Package		a)	MCP6541T-I/LT:	5LD SC-70, Tape and Reel.
	Range		b)	MCP6541T-I/OT:	5LD SOT-23. Tape and Reel.
Device:	MCP6541: Single Comparator MCP6541T: Single Comparator (Tape and Reel)		c)	MCP6541-I/P:	8LD PDIP.
	(SC-70, SOT-23, SOIC, MSOP)		a)	MCP6542-I/MS:	8LD MSOP.
	MCP6542: Dual Comparator		b)	MCP6542T-I/MS:	8LD MSOP, Tape and Reel.
	MCP6542T: Dual Comparator (Tape and Reel for SOIC and MSOP) MCP6543: Single Comparator with CS		C)	MCP6542-I/P:	8LD PDIP.
	MCP6543T: Single Comparator with CS		a)	MCP6543-I/SN:	8LD SOIC.
	(Tape and Reel for SOIC and MSOP) MCP6544: Quad Comparator		b)	MCP6543T-I/SN:	8LD SOIC, Tape and Reel.
	MCP6544T: Quad Comparator (Tape and Reel for SOIC and TSSOP)		C)	MCP6543-I/P:	8LD PDIP.
Temperature Range:	$= -40^{\circ}$ C to +85°C		a)	MCP6544T-I/SL:	14LD SOIC, Tape and Reel.
Temperature Range.			b)	MCP6544T-I/SL:	14LD SOIC, Tape and Reel.
Package:	LT = Plastic Package (SC-70), 5-lead OT = Plastic Small Outline Transistor (SOT-23), 5-lead MS = Plastic MSOP, 8-lead P = Plastic DIP (300 mil Body), 8-lead, 14-lead SN = Plastic SOIC (150 mil Body), 8-lead SL = Plastic SOIC (150 mil Body), 14-lead (MCP6544) ST = Plastic TSSOP (4.4mm Body), 14-lead (MCP6544)			MCP6544-I/P:	14LD PDIP.

#### Sales and Support

#### **Data Sheets**

Products supported by a preliminary Data Sheet may have an errata sheet describing minor operational differences and recommended workarounds. To determine if an errata sheet exists for a particular device, please contact one of the following:

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- The Microchip Worldwide Site (www.microchip.com)

Please specify which device, revision of silicon and Data Sheet (include Literature #) you are using.

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## MCP6541/2/3/4

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- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
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- Microchip is willing to work with the customer who is concerned about the integrity of their code.
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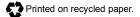
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